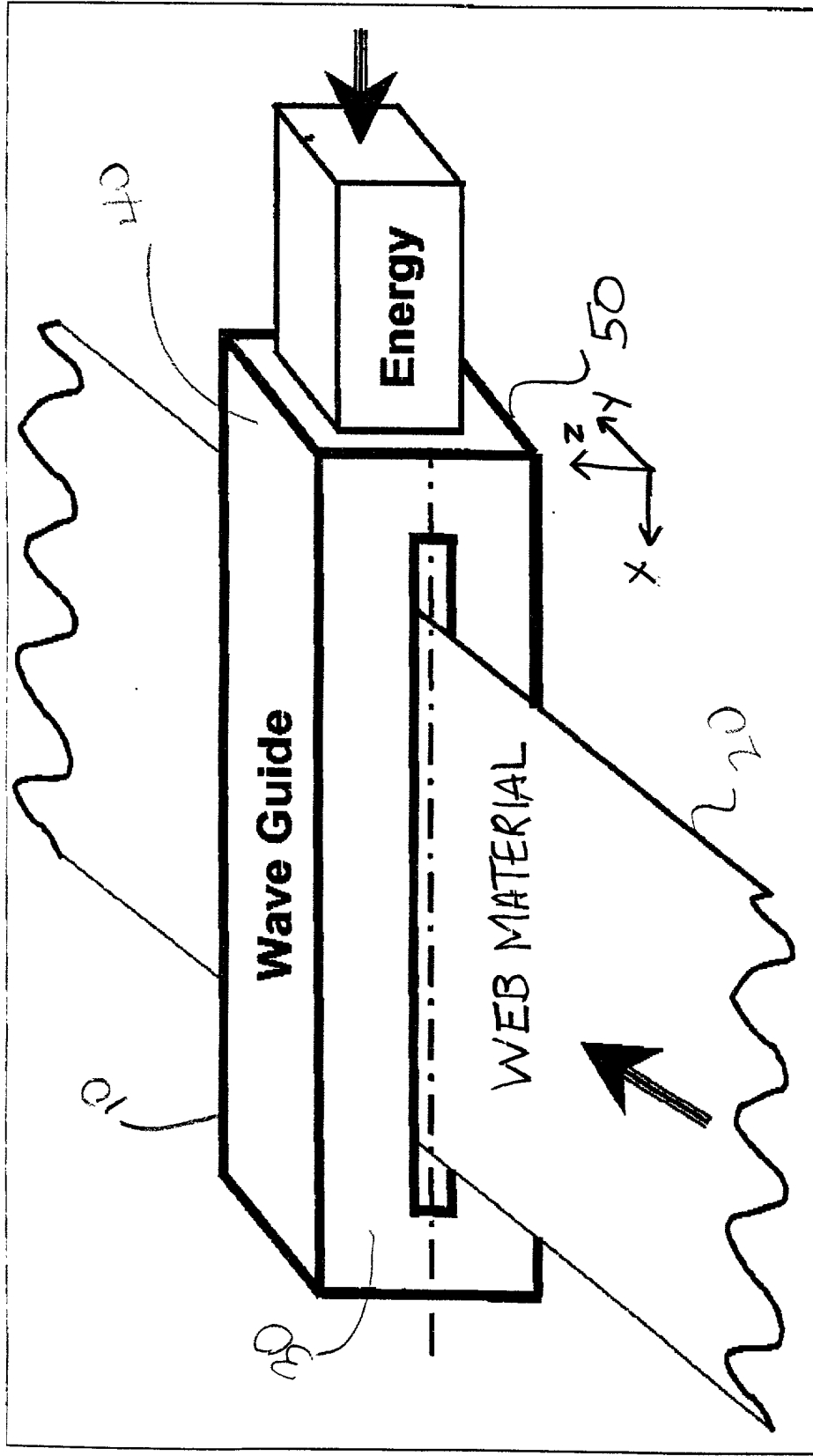


Centerline Slot Implementation



Non-Centerline Slot

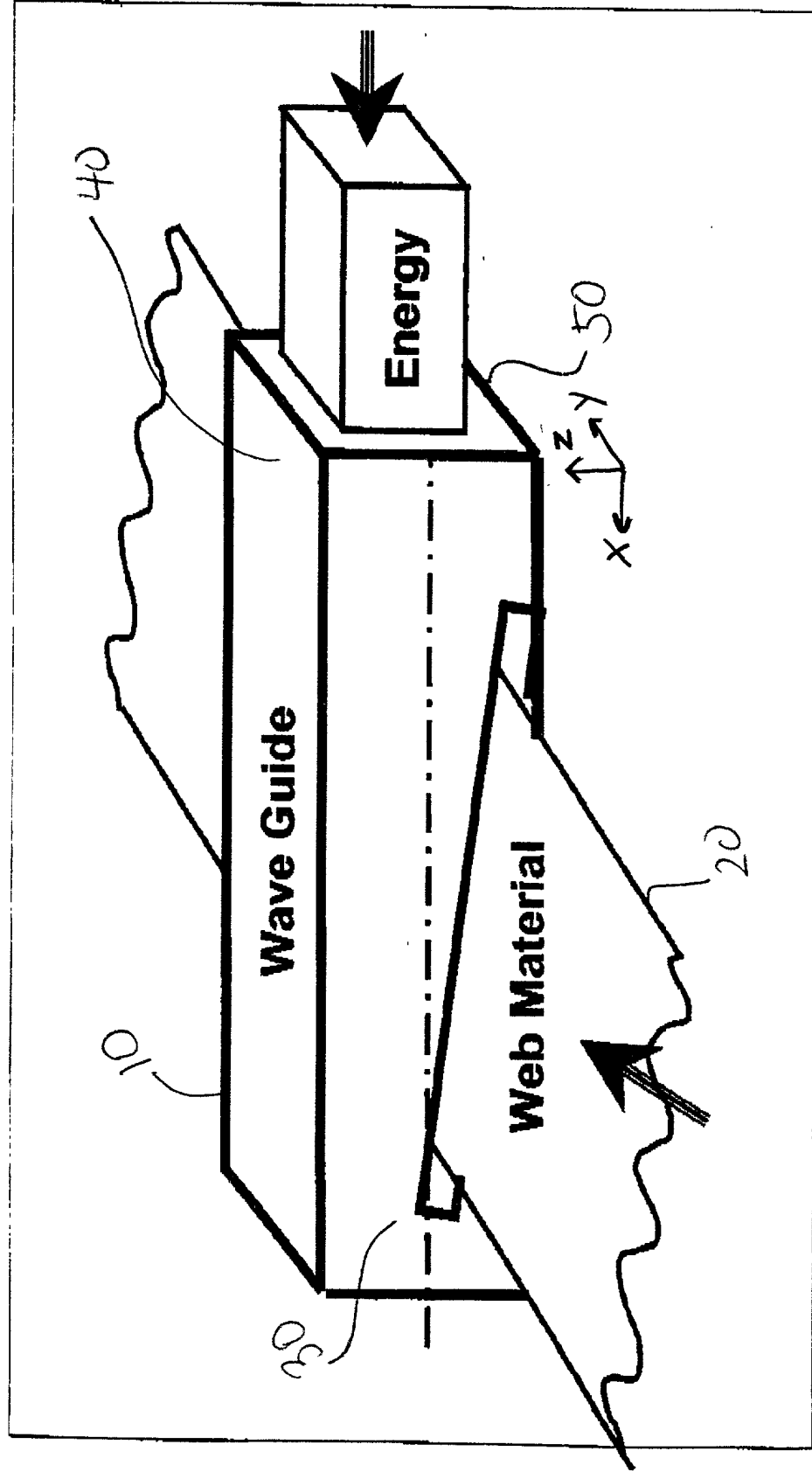
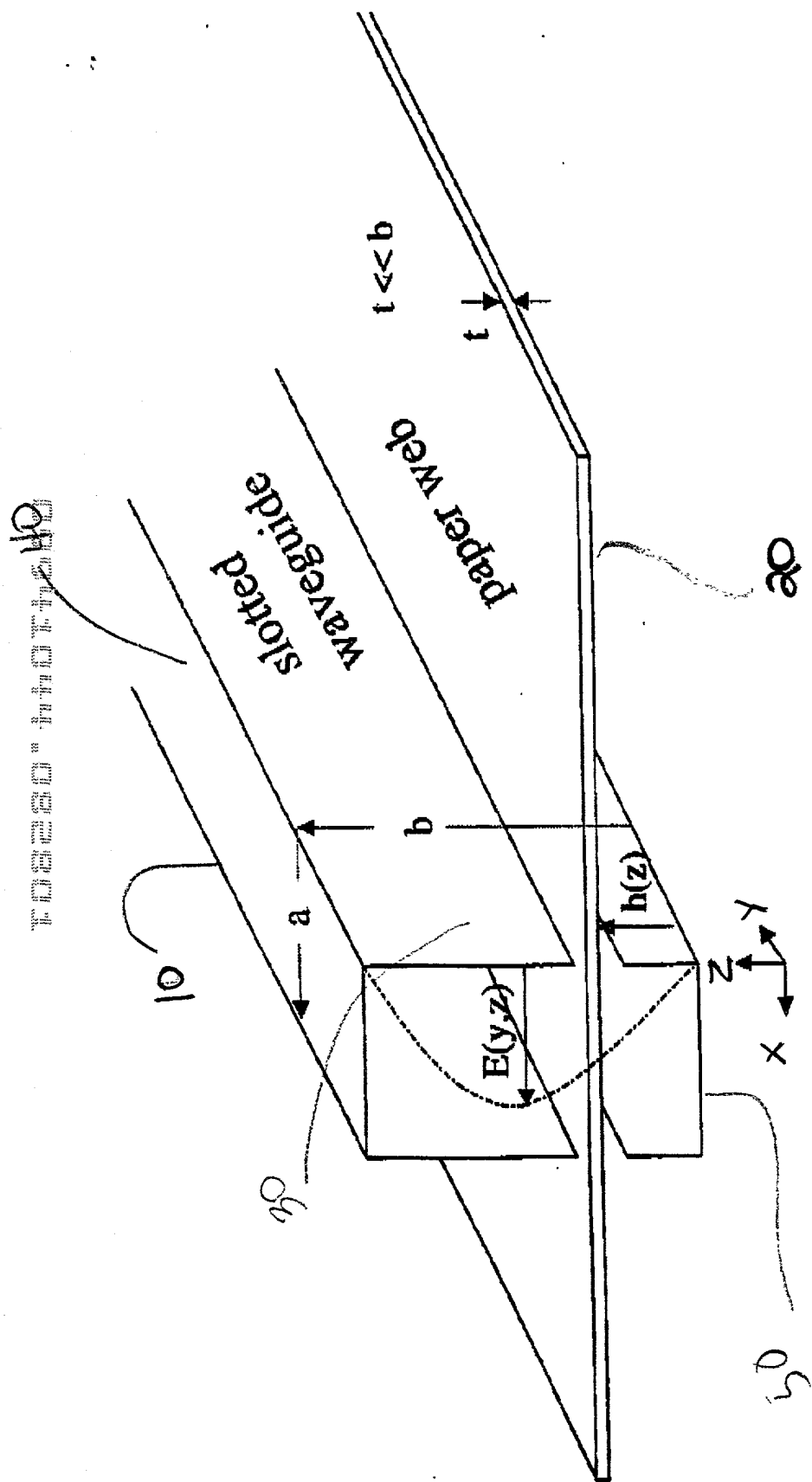
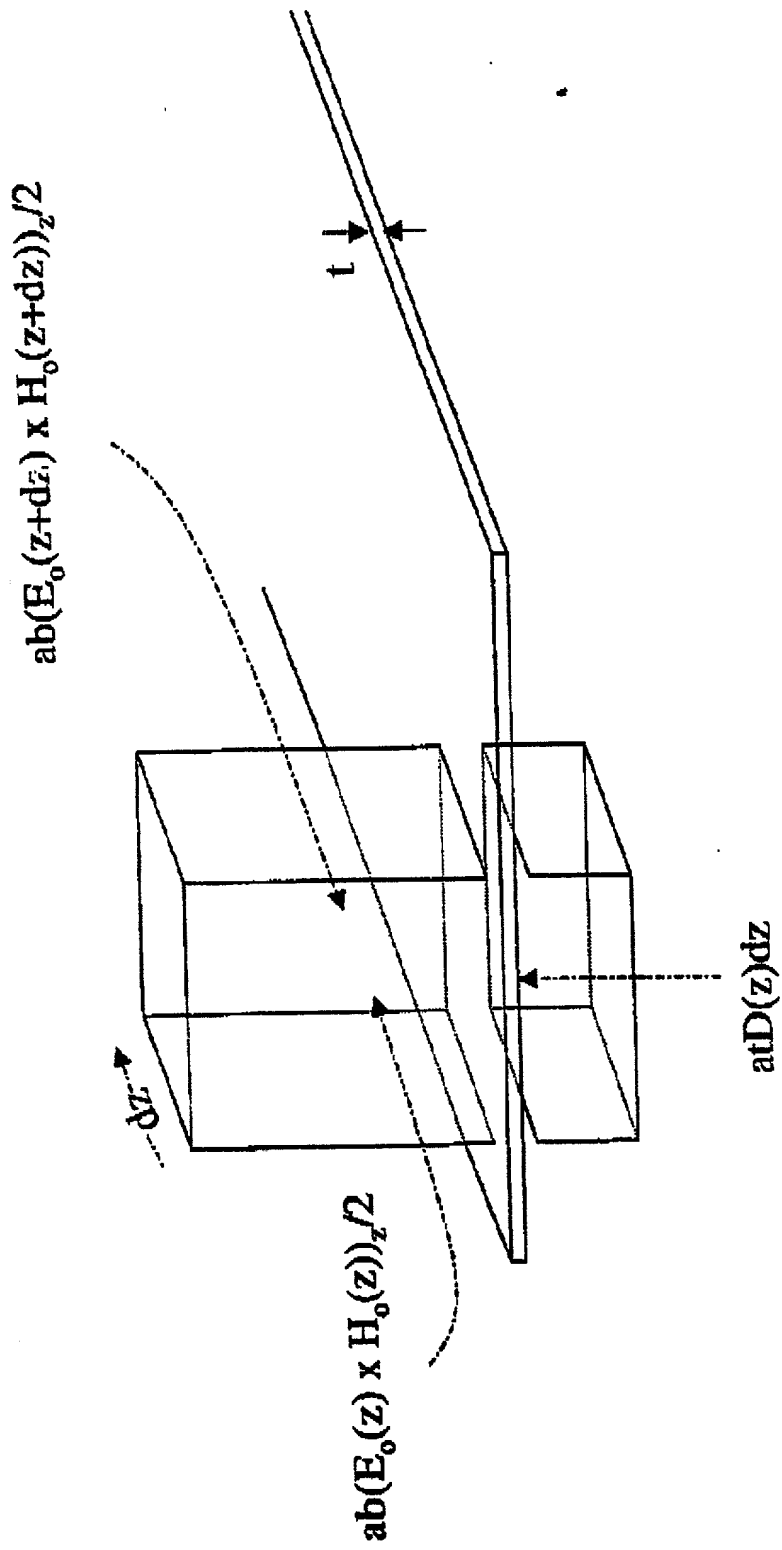


Fig. 2 PRIOR ART



Parameters for Paper Drying in a Waveguide

Figure 3



Schematic for energy balance on an infinitesimal guide section

Figure 4

Effect of using a linear slot profile

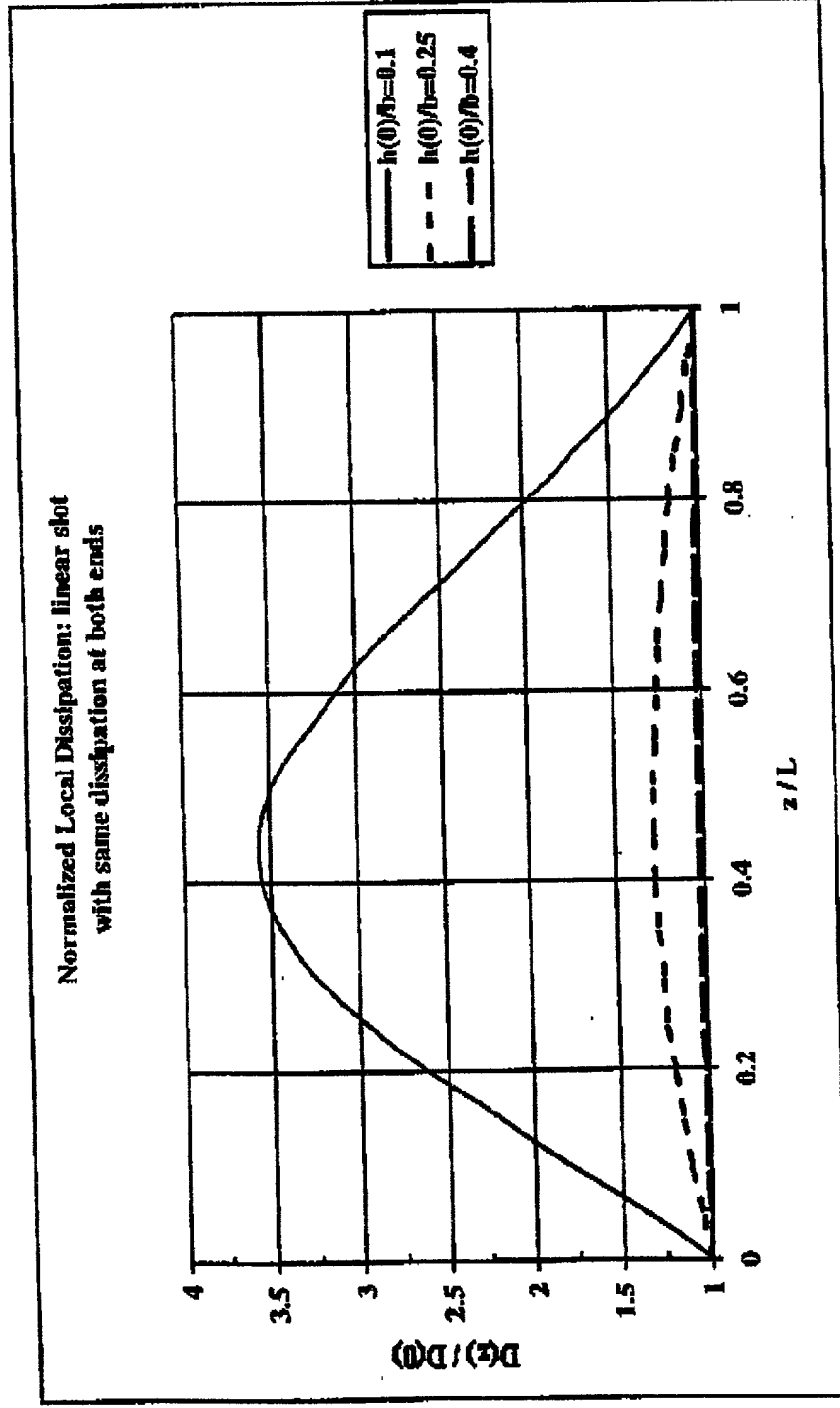
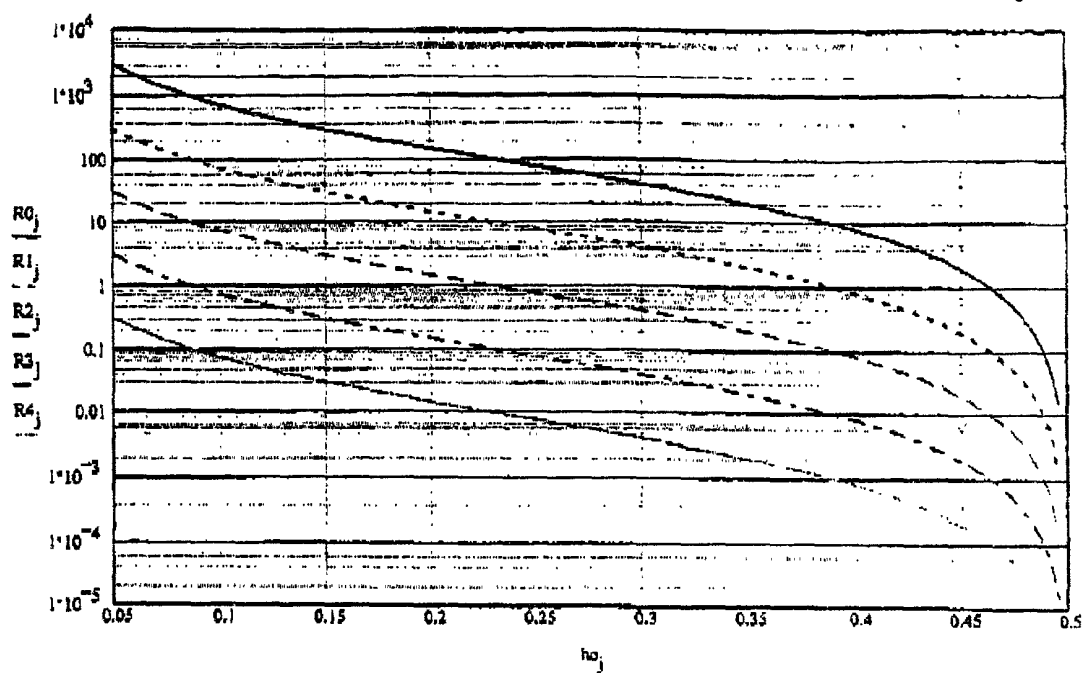


Figure 5

Linear Slot Dissipation Profile as a Function of Starting Slot Height



These are plots of the range of curved-slot-compensated waveguide as a function of h_o/b , the ratio of the starting slot height to the guide breadth. Curves are drawn for different values of ϵr^t in meters. The values of ϵr^t plotted are listed below. The curves drop to lower values as ϵr^t increases

Figure 6

$b = 0.072$ guide breadth in m

$f = 2.45 \cdot 10^9$ frequency in Hz

$\sin(\pi \cdot \min)^2 = 0.024$

$$\epsilon r^t = \begin{bmatrix} 5 \cdot 10^{-6} \\ 5 \cdot 10^{-5} \\ 5 \cdot 10^{-4} \\ 5 \cdot 10^{-3} \\ 0.05 \end{bmatrix}$$

Now calculate the shape of a slot curve for a given $\epsilon''t$ and h/b

$\epsilon''t := 10^{-4}$ enter web imaginary dielectric constant times thickness in meters

$N := 1000$ enter number of data points in a slot curve plot

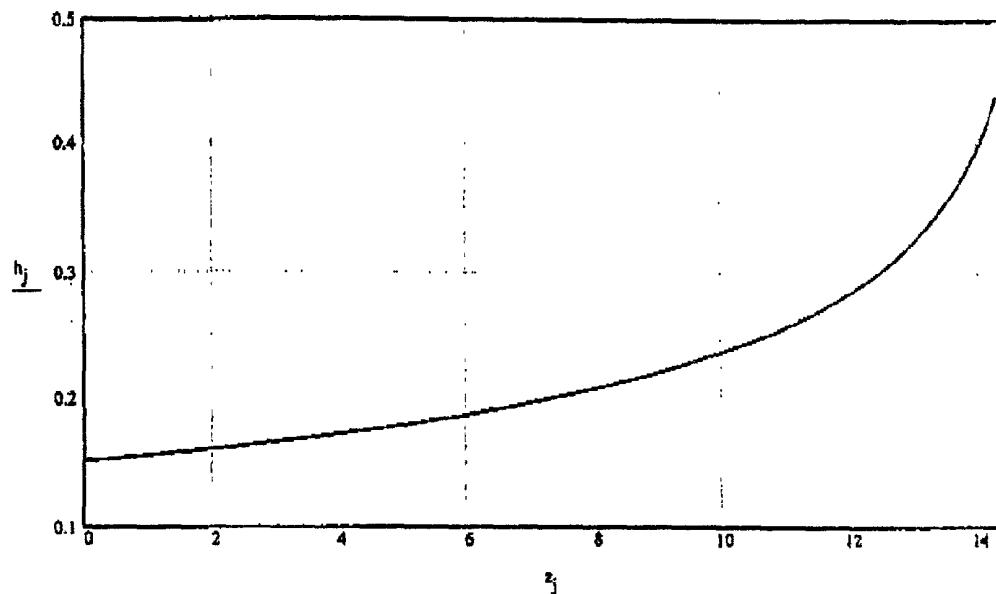
$j := 0..N-1$ set up iteration parameter for range plots

$h_{min} := .15$ enter starting ratio of h/b

$$z_{max} := \frac{b \cdot \left(\frac{1}{\sin(\pi \cdot h_{min})^2} - 1 \right)}{2 \cdot \omega \cdot Z_0 \cdot \epsilon''t} \quad \text{calculate maximum value of compensated } z$$

$$z_j := .99 \cdot z_{max} \cdot \frac{j}{N-1} \quad \text{generate values for slot height plots}$$

$$h_j := \left(\frac{1}{\pi} \right) \cdot \arcsin \left[\left(\frac{1}{\sin(\pi \cdot h_{min})^2} - 2 \cdot \omega \cdot Z_0 \cdot \epsilon''t \cdot \frac{z_j}{b} \right)^{\frac{1}{2}} \right] \quad \text{calculate slot height values normalized to } b \text{ as a function of } z$$



This is a plot of height of the slot divided by the guide breadth as a function of guide length in meters

Figure 7

$\epsilon''t = 1 \cdot 10^{-4}$ web imaginary dielectric constant times thickness (m)

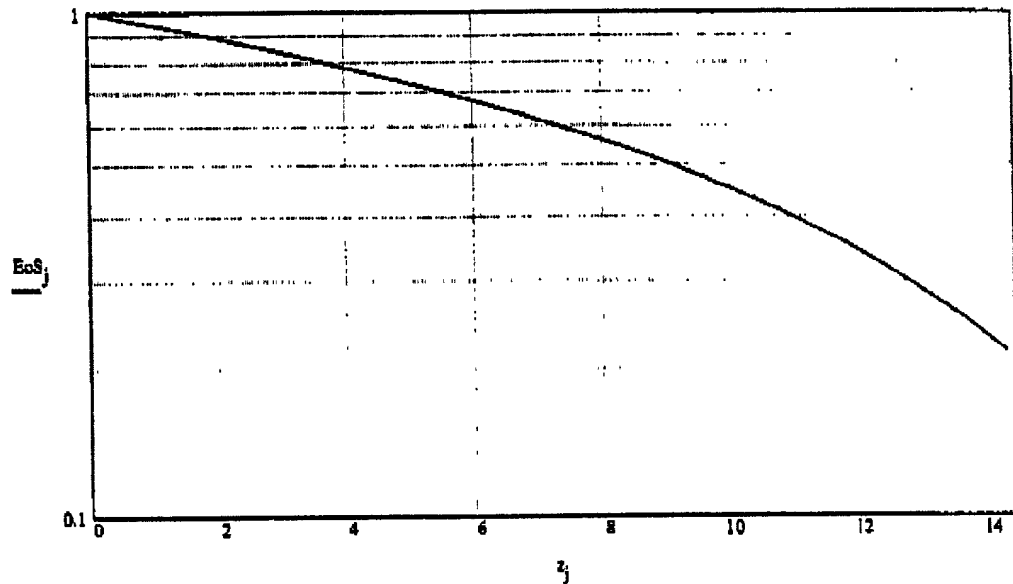
$h_{min} = 0.15$ initial h/b

$z_{max} = 14.443$ range of compensation in meters

Calculate the ratio of the E field intensity at the guide center to its initial value as a function of z for the same parameters as in the slot shape curve just above.

$$EoS_z := \left(1 - 2 \cdot \omega \cdot Z \cdot \epsilon_0 \frac{\epsilon_{rt}}{b} \cdot z_j \cdot \sin(\pi \cdot h_{omin})^2 \right)$$

calculate the ratio of Eo squared to Eoo squared as a function of z



This is a plot of the relative center guide field intensity versus guide length for an IMS optimum compensated slotted waveguide. The z axis is in meter and y axis is the intensity is ratioed to its value at z=0.

Figure 8

$$\epsilon_{rt} = 1 \cdot 10^{-4}$$

web imaginary dielectric
constant times thickness (m)

$$h_{omin} = 0.15$$

initial h/b

$$z_{max} = 14.443$$

range of compensation in meters

M := 4 enter number of web runs

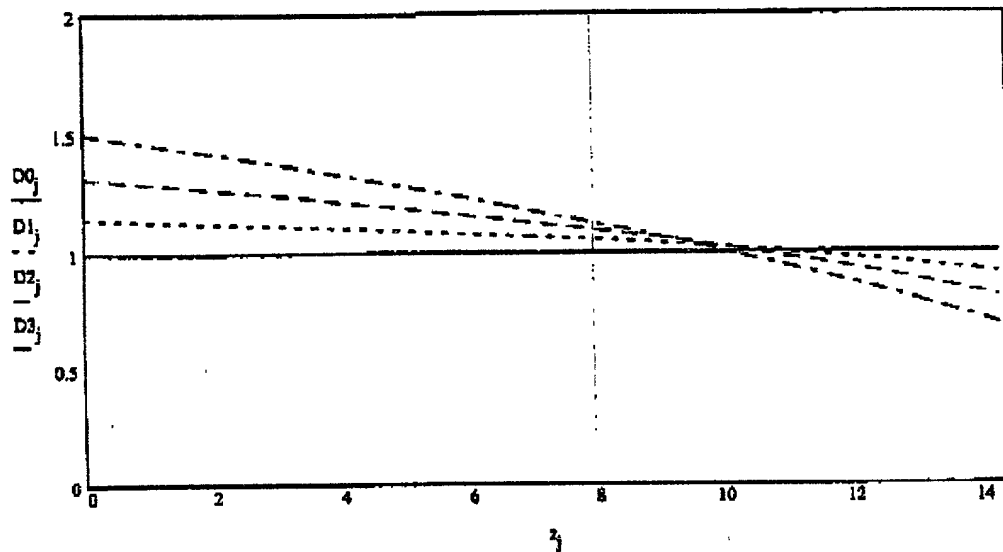
R := 1.5 enter maximum ratio of ert operation to ert designed

m := 0..M - 1 iteration parameter

$r_m := R^{\frac{m}{M-1}}$ calculate the values of the ratio of the actual ert to the designed ert

$$D0_j := r_0 \cdot \left(1 - 2 \cdot \omega \cdot Z \cdot \epsilon_0 \cdot \frac{\text{ert}}{b} \cdot z_j \cdot \sin(\pi \cdot \text{homin})^2 \right)^{r_0 - 1} \quad D1_j := r_1 \cdot \left(1 - 2 \cdot \omega \cdot Z \cdot \epsilon_0 \cdot \frac{\text{ert}}{b} \cdot z_j \cdot \sin(\pi \cdot \text{homin})^2 \right)^{r_1 - 1}$$

$$D2_j := r_2 \cdot \left(1 - 2 \cdot \omega \cdot Z \cdot \epsilon_0 \cdot \frac{\text{ert}}{b} \cdot z_j \cdot \sin(\pi \cdot \text{homin})^2 \right)^{r_2 - 1} \quad D3_j := r_3 \cdot \left(1 - 2 \cdot \omega \cdot Z \cdot \epsilon_0 \cdot \frac{\text{ert}}{b} \cdot z_j \cdot \sin(\pi \cdot \text{homin})^2 \right)^{r_3 - 1}$$



These are plots of the web heat dissipation relative to the heat dissipation at $z=0$ in the designed waveguide as a function of waveguide length in meters. Different curves have different ratios of ert operating to ert designed. The actual ratios are listed below as r

Figure 9

$\text{ert} = 1 \cdot 10^{-4}$ designed web imaginary dielectric constant times thickness (m)

$z_{\text{max}} = 14.443$ range of compensation in meters

$\text{homin} = 0.15$ initial h/b

$$r = \begin{bmatrix} 1 \\ 1.145 \\ 1.31 \\ 1.5 \end{bmatrix}$$

Two Serpentine Microwave Applicator Configurations: (a) Short at Termination End; (b) Dummy Load at Termination End.

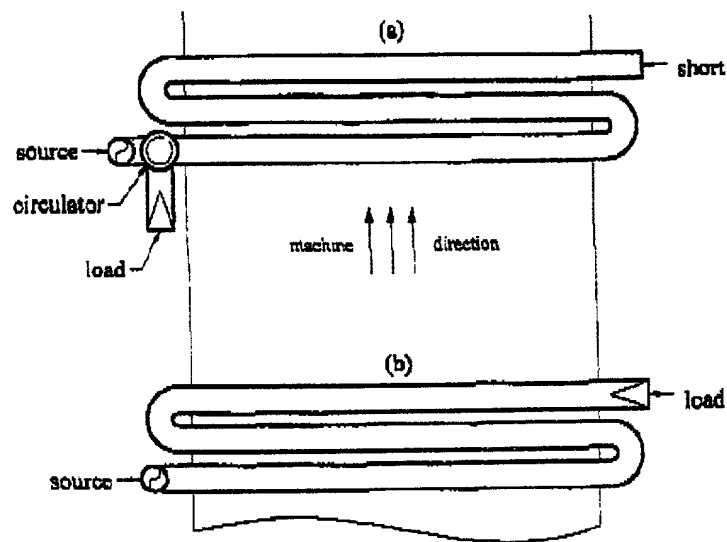


FIGURE 10

Definition of Slot (and Paper)

Location within the Waveguide. The cross-machine coordinate is z and $h(z)$ is the local elevation of the slot above the bottom of the waveguide. The overall active cross-machine length is L .

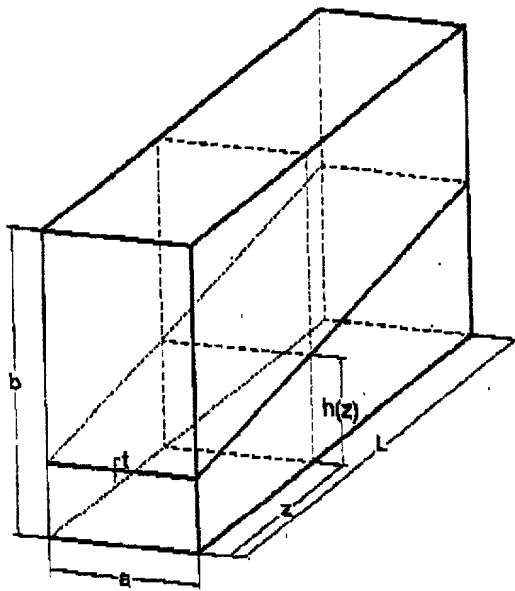
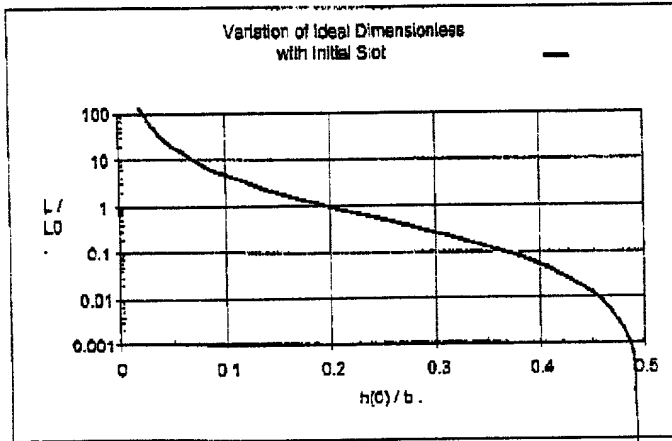


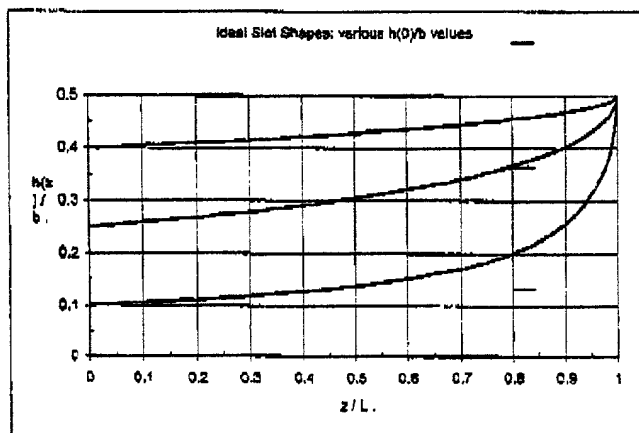
FIGURE 11



Ideal Dimensionless Length vs. Initial Slot Height.

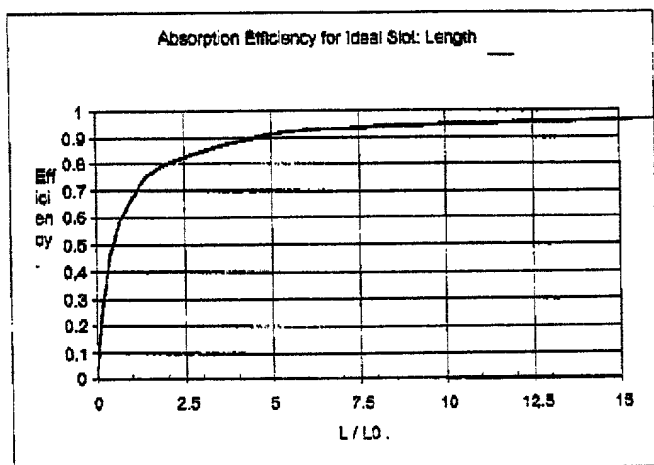
FIGURE 12

10630-4404-400



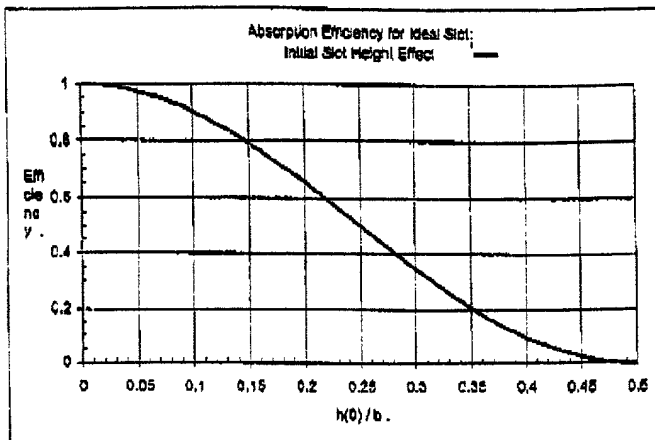
Ideal Slot Shapes for $h(0)/b = 0.1, 0.25, 0.4$.

FIGURE 13



Efficiency vs. Ideal Dimensionless Length.

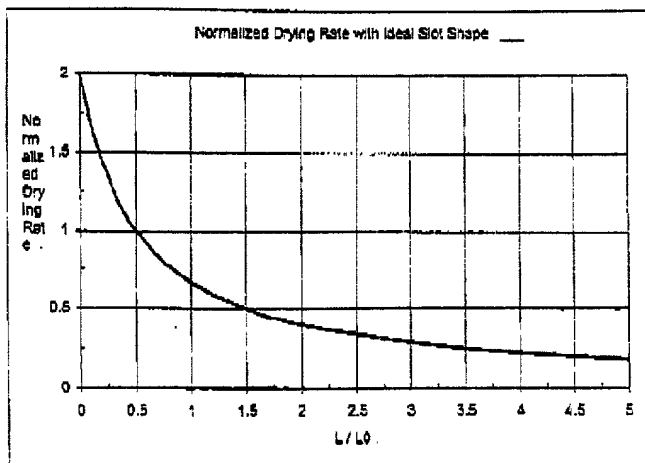
FIGURE 14



Efficiency (at Ideal Length) vs. Initial Height.

FIGURE 15

Normalized Drying Rate with Ideal Slot Shape



Normalized Drying Rate for Ideal Length.

FIGURE 16

The slot height profile, $h(z)$, which gives uniform drying depends on the paper basis weight and its moisture content, $\epsilon_r''t$.

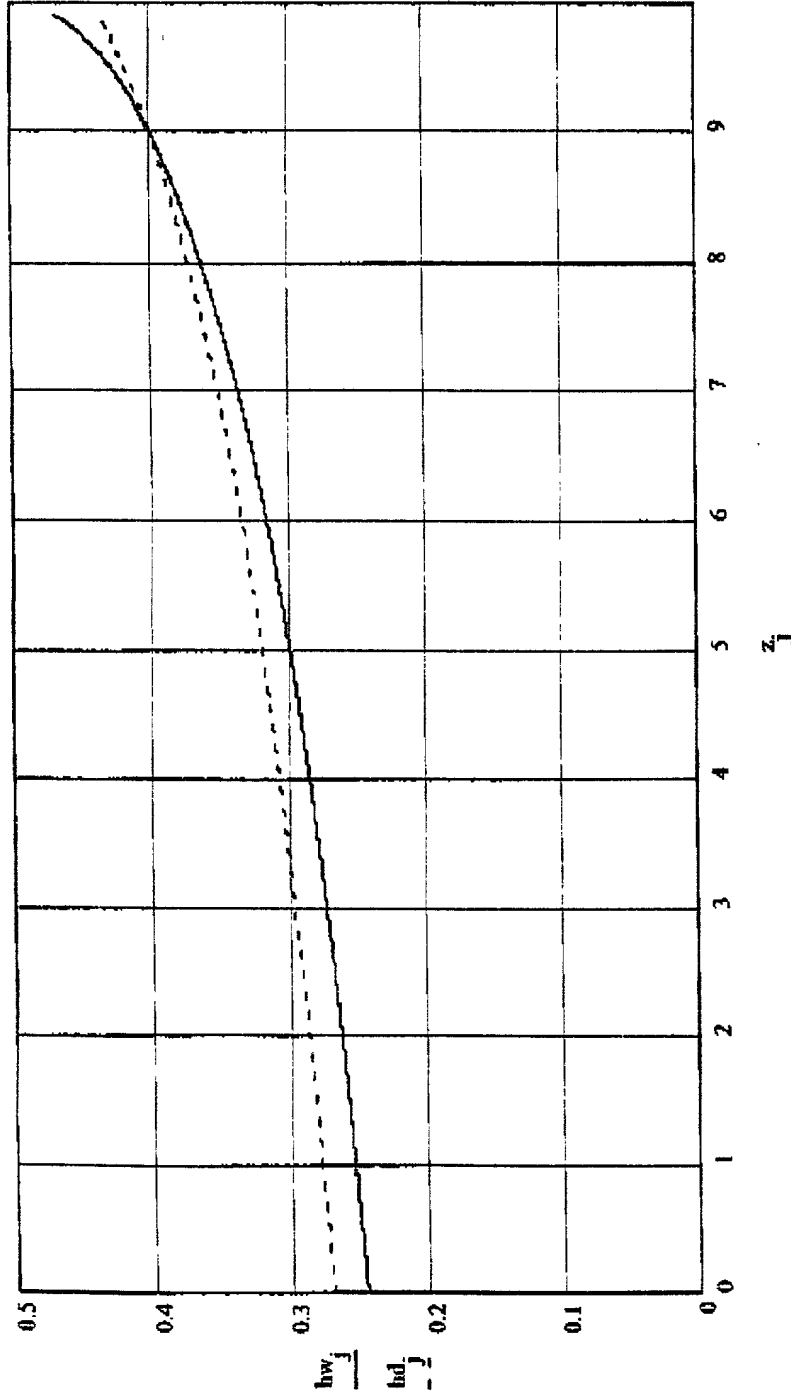
The optimal slot profile is

$$h(z) = (b/\pi)\sin^{-1}[(1/\sin^2(\pi h_0/b) - 2Z\omega\epsilon_0\epsilon_r''tZ/b)^{-1/2}]$$

where h_0 represents the slot height at the source side of the web and z is the distance along the waveguide (CD).

Fig. 17

Optimal Slot Profiles



Plots of the optimal slot height divided by the waveguide height as a function of distance in meters from a microwave source at 2.45 GHz in an S-band waveguide. The solid line is designed for a 200 g/m² board at 10% moisture, whereas the dotted line is for 7% moisture.

Fig. 18

FEED METHOD

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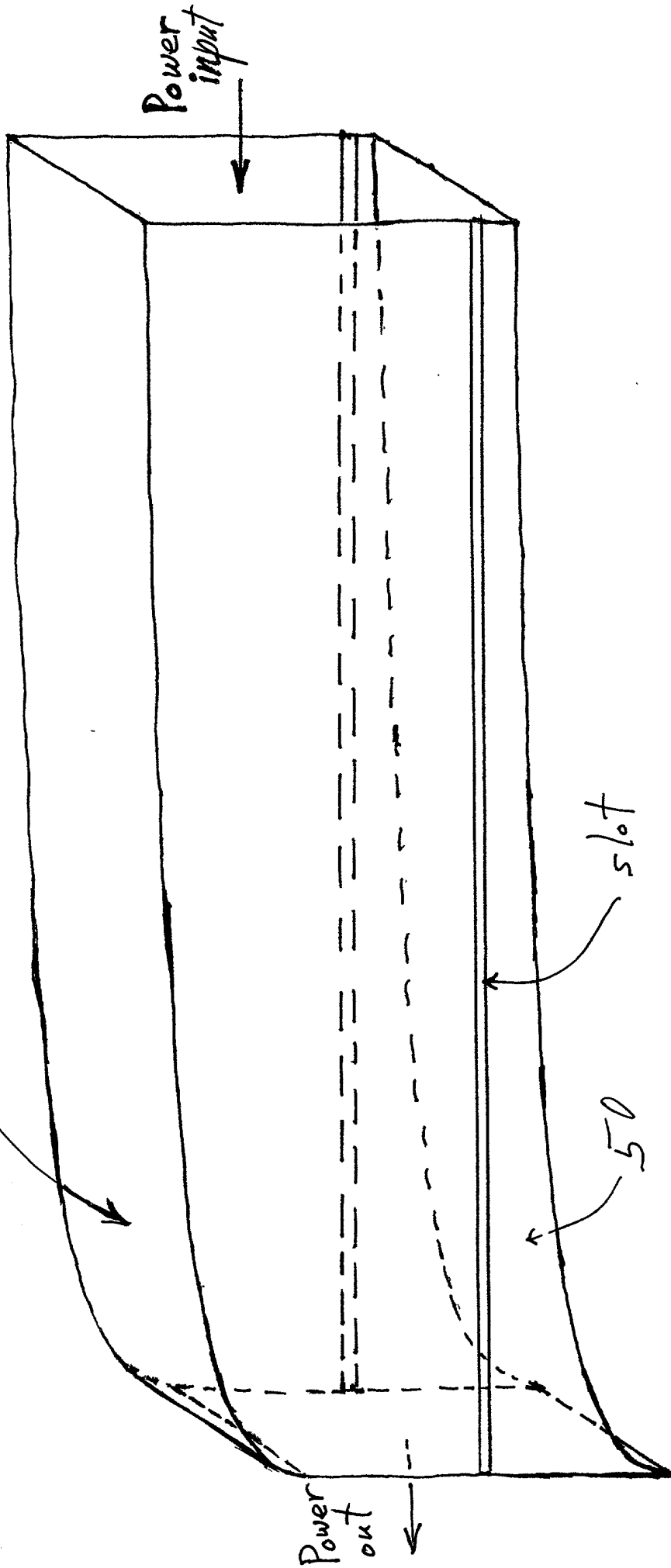
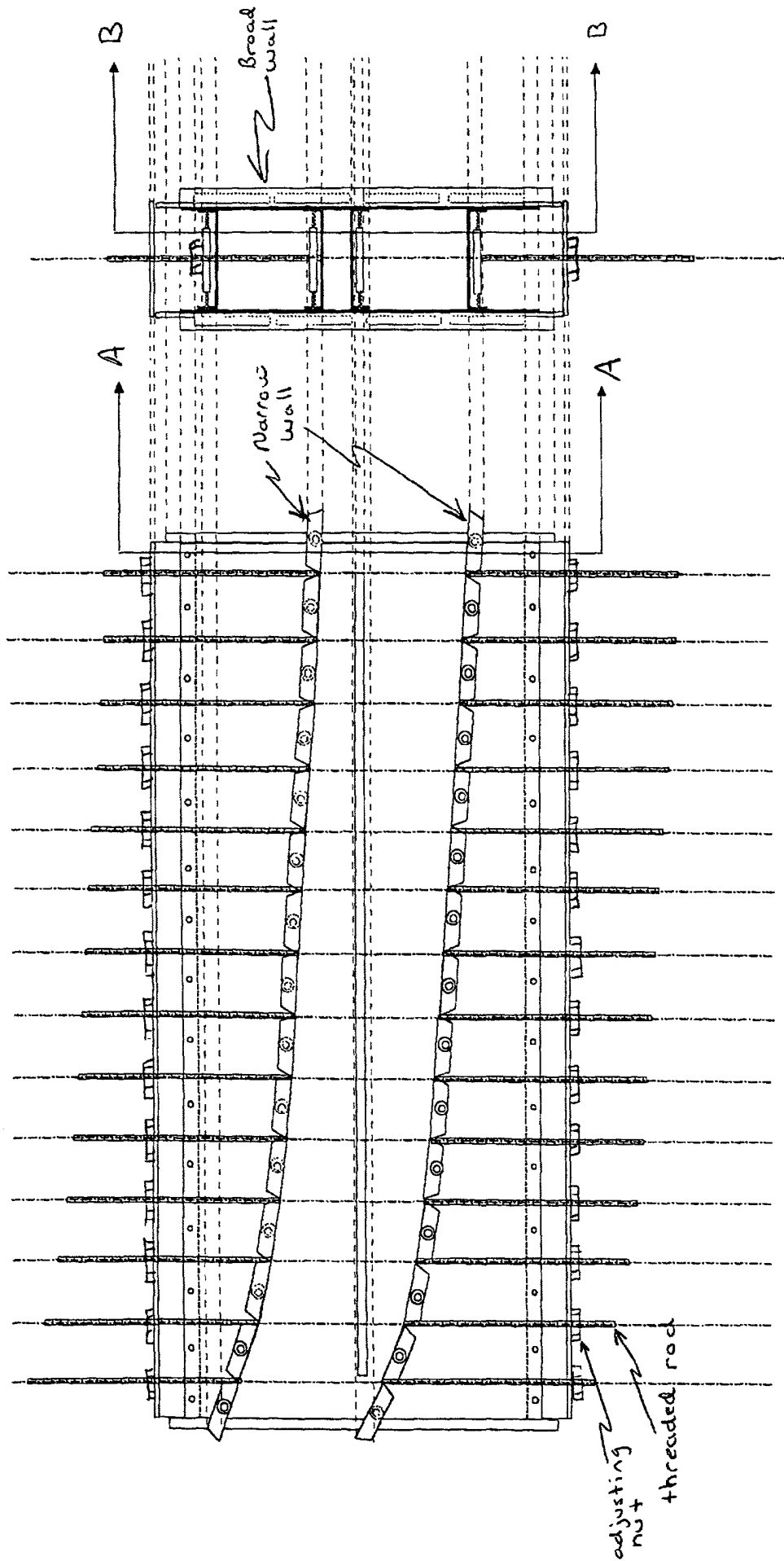


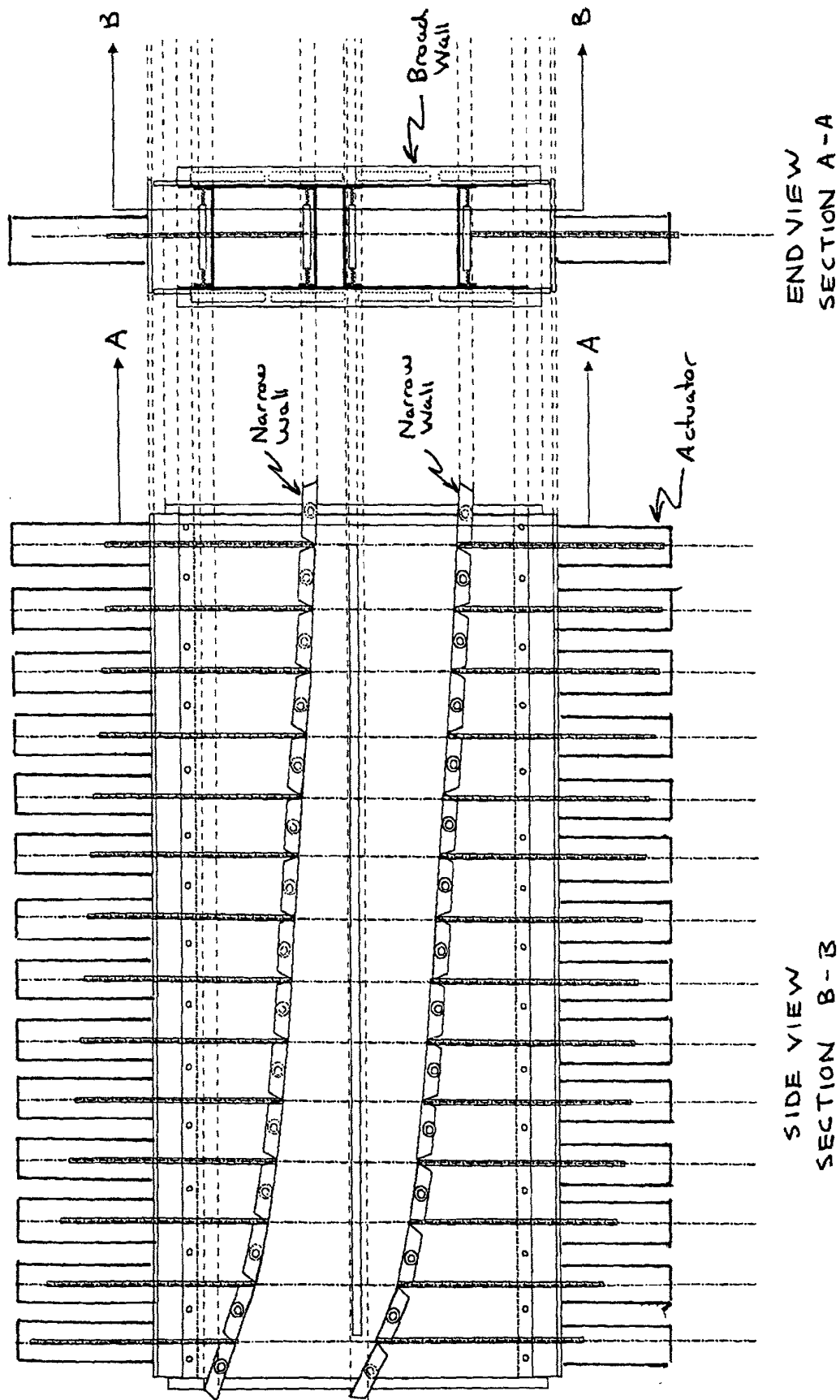
FIGURE 19



END VIEW
SECTION A-A

SIDE VIEW - SECTION B-B
Manually Adjusted Variable Waveguide

Fig. 20



Automatically Adjusted Variable Waveguide

FIG 21